

LINEAR INTEGRATED CIRCUITS



PREAMPLIFIER WITH ALC FOR C<sub>7</sub>O<sub>2</sub> CASSETTE RECORDERS

- EXCELLENT VERSATILITY IN USE (V<sub>S</sub> from 4 to 20V)
- HIGH OPEN LOOP GAIN
- LOW DISTORTION
- LOW NOISE
- LARGE AUTOMATIC LEVEL CONTROL RANGE
- STEREO MATCHING BETTER THAN 3 dB (matched pair)

The TDA 2054M is a monolithic integrated circuit in a 16-lead dual in-line plastic package.

The functions incorporated are:

- low noise preamplifier
- automatic level control system (ALC)
- high gain equalization amplifier

It is intended as preamplifier in tape and cassette recorders and players (C<sub>7</sub>O<sub>2</sub>), dictaphones, compressor and expander in telephonic equipments, Hi-Fi preamplifiers and in wire diffusion receivers; for stereo applications the ALC matching is better than 3 dB.

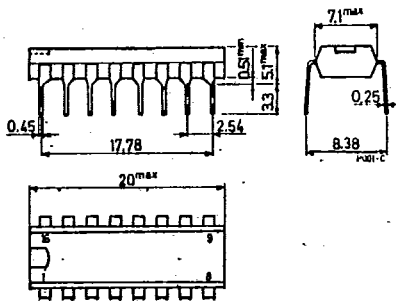
ABSOLUTE MAXIMUM RATINGS

V <sub>S</sub>	Supply voltage	20	V
P <sub>tot</sub>	Total power dissipation at T <sub>amb</sub> = 50°C	500	mW
T <sub>stg</sub> , T <sub>j</sub>	Storage and junction temperature	-40 to 150	°C

ORDERING NUMBERS: TDA 2054M mono applications  
2 TDA 2054M stereo applications

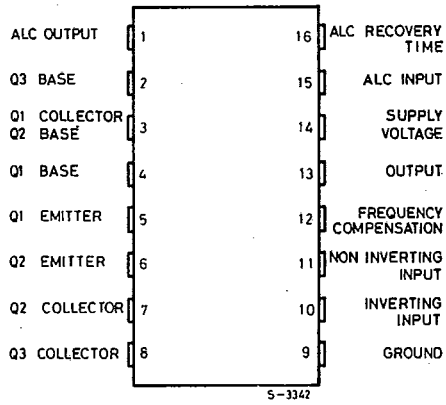
MECHANICAL DATA

Dimensions in mm

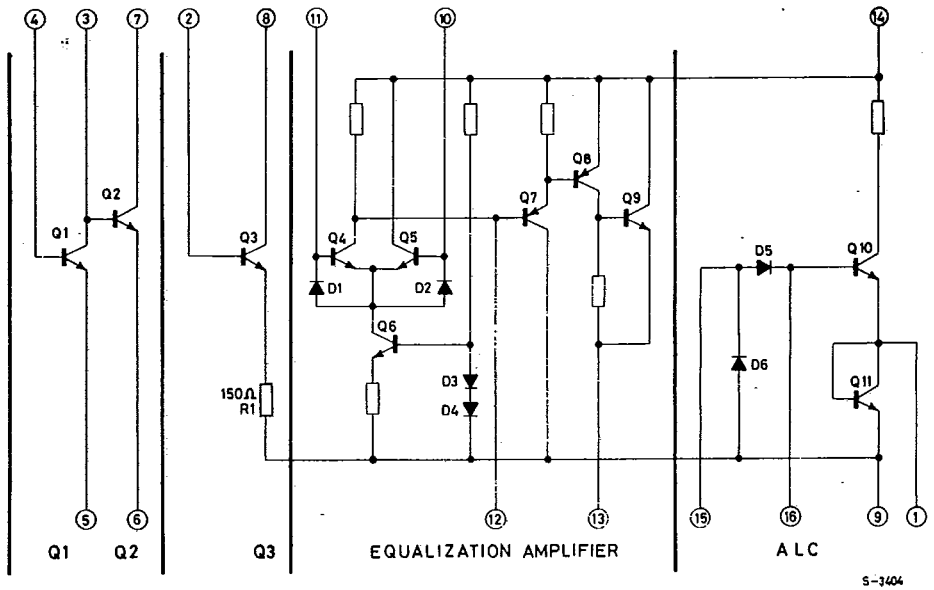




CONNECTION DIAGRAM

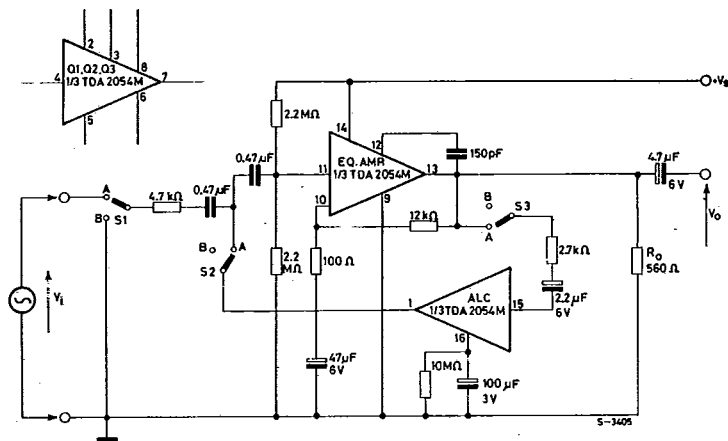


SCHEMATIC DIAGRAM





TEST CIRCUIT



THERMAL DATA

$R_{th J-amb}$	Thermal resistance junction-ambient	max	200	°C/W
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ELECTRICAL CHARACTERISTICS (Refer to the test circuit,  $T_{amb} = 25^{\circ}C$ )

Parameter	Test conditions	Min.	Typ.	Max.	Unit	
$V_s$	Supply voltage	4		20	V	
$I_d$	Quiescent drain current		10		mA	
$h_{FE}$	DC current gain (Q1, Q2, Q3)	$I_c = 0.1 \text{ mA}$ $V_{CE} = 5V$	300	500	-	
$e_N$	Input noise voltage (Q1, Q2, Q3)	$I_c = 0.1 \text{ mA}$ $V_{CE} = 5V$ $f = 1 \text{ KHz}$		2	$\frac{nV}{\sqrt{Hz}}$	
$i_N$	Input noise current (Q1, Q2, Q3)		0.5		$\frac{pA}{\sqrt{Hz}}$	
NF	Noise figure (Q1, Q2, Q3)	$I_c = 0.1 \text{ mA}$ $V_{CE} = 5V$ $R_q = 4.7 \text{ K}\Omega$ $B (-3 \text{ dB}) = 20 \text{ to } 10000 \text{ Hz}$		0.5	4	dB
$G_v$	Open loop voltage gain (for equalization amplifier)	$V_s = 9V$ $f = 1 \text{ KHz}$	60		dB	
$V_o$	Output voltage with A.L.C.	$V_s = 9V$ $f = 1 \text{ KHz}$ $V_I = 100 \text{ mV}$ $S1 = S2 = S3 \text{ at A}$	0.6		V	
$e_N$	Equivalent input noise voltage (for equalization amplifier pin 11)	$V_s = 9V$ $G_v = 40 \text{ dB}$ $B (-3 \text{ dB}) = 20 \text{ to } 20000 \text{ Hz}$ $S1 \text{ at B}$	1.3		$\mu V$	
$R_1$	Q3 emitter resistance	105	150	195	$\Omega$	



Fig. 1 - Equivalent input spot voltage and noise current vs. bias current (transistors Q1, Q2, Q3)

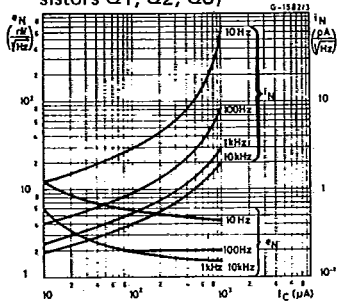


Fig. 2 - Equivalent input noise current vs. frequency (transistors Q1, Q2, Q3)

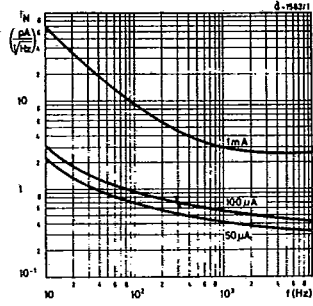


Fig. 3 - Equivalent input noise voltage vs. frequency (transistors Q1, Q2, Q3)

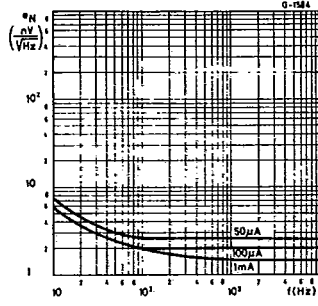


Fig. 4 - Noise figure vs. bias current (transistors Q1, Q2, Q3)

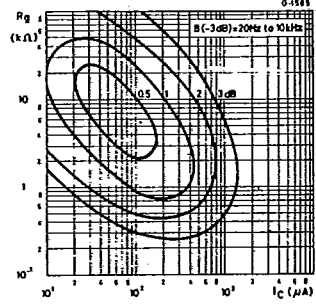


Fig. 5 - Optimum source resistance and minimum NF vs. bias current (transistors Q1, Q2, Q3)

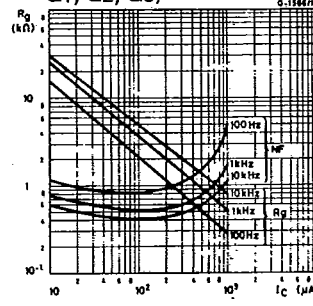


Fig. 6 - Current gain vs. collector current (transistors Q1, Q2, Q3)

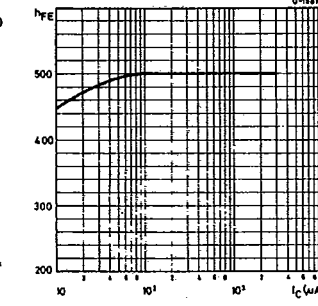


Fig. 7 - Open loop gain vs. frequency (equalization amplifier)

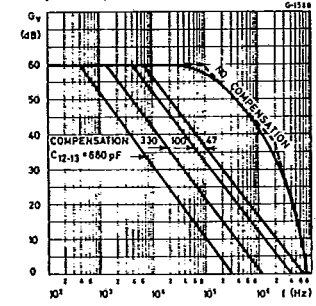


Fig. 8 - Open loop phase response vs. frequency (equalization amplifier)

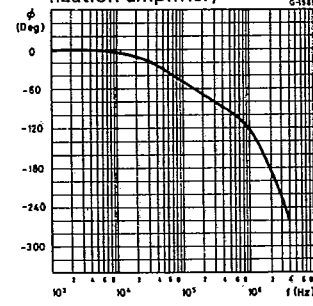
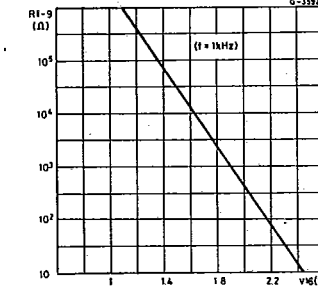


Fig. 9 - Dynamic resistance R1-9 vs. ALC voltage V16





APPLICATION INFORMATION

Fig. 9 - Application circuit for C<sub>r</sub>O<sub>2</sub> cassette player and recorder

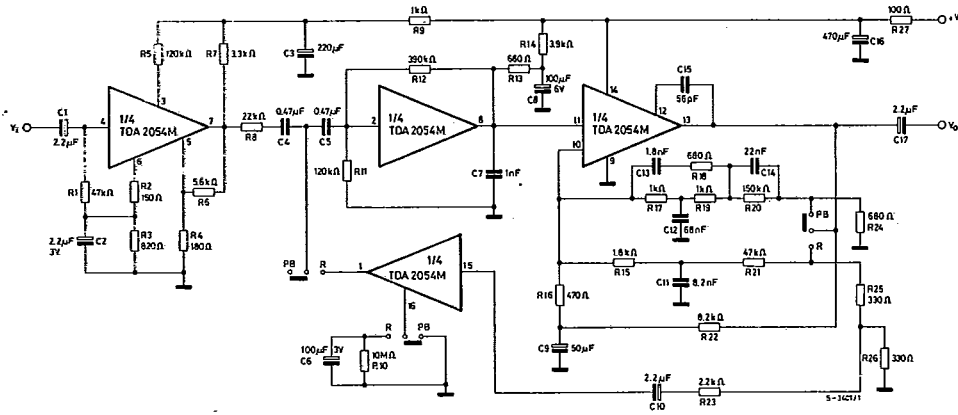
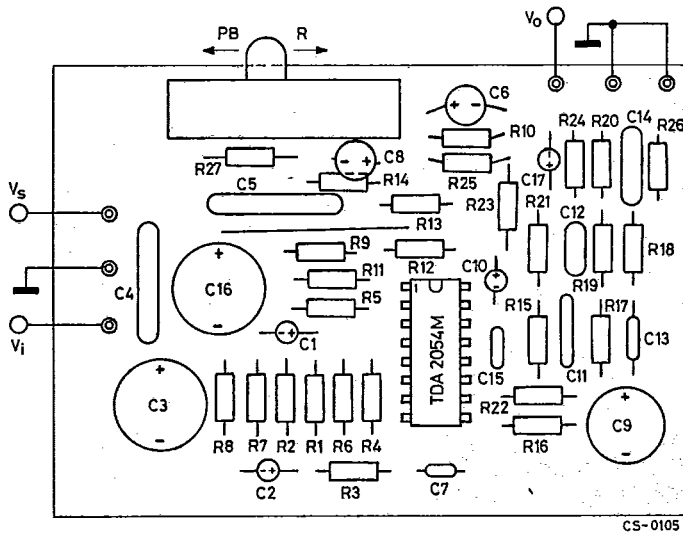


Fig. 10 - P.C. board and component layout for the circuit of Fig. 9 (1:1 scale)





TYPICAL PERFORMANCE OF CIRCUIT IN FIG. 9 ( $T_{amb} = 25^{\circ}C$ ,  $V_s = 9V$ )

Parameter	Test conditions	Min.	Typ.	Max.	Unit
<b>PLAYBACK</b>					
$G_v$	Voltage gain (open loop)		134		dB
$G_v$	Voltage gain (closed loop)		60		dB
$Z_i$	Input impedance	$f = 100 \text{ Hz}$ $f = 1 \text{ KHz}$ $f = 10 \text{ KHz}$	10 41 43		$K\Omega$ $K\Omega$ $K\Omega$
$Z_o$	Output impedance	$f = 1 \text{ KHz}$	12	35	$\Omega$
B	Frequency response		see fig. 11		
d	Distortion	$V_o = 1V$ $f = 1 \text{ KHz}$	0.2		%
	Output background noise	$Z_g = 300\Omega + 120 \text{ mH}$ (DIN 45405)	1.5		mV
***	Output weighted background noise		1		mV
$\frac{S+N}{N}$	Signal to noise ratio	$V_o = 1.5V$ $Z_g = 300\Omega + 120 \text{ mH}$	60		dB
$t_{on}^*$	Switch-on time	$V_o = 1V$	500		ms

**RECORDING**

$G_v$	Voltage gain (open loop)	$f = 20 \text{ to } 20000 \text{ Hz}$	134		dB
$G_v$	Voltage gain (closed loop)	$f = 1 \text{ KHz}$	72		dB
B	Frequency response		see fig. 13		
d	Distortion with ALC	$V_o = 1V$ $f = 10 \text{ KHz}$	0.5		%
ALC	Automatic level control range (for 3 dB of output voltage variation)	$V_i \leq 40 \text{ mV}$ $f = 10 \text{ KHz}$	54		dB
$V_o$	Output voltage before clipping without ALC	$f = 1 \text{ KHz}$	3		V
$V_o$	Output voltage with ALC	$V_i = 30 \text{ mV}$ $f = 1 \text{ KHz}$	1.1		V
$t_l^*$	Limiting time (see fig. 17)	$\Delta V_i = +40 \text{ dB}$ $f = 1 \text{ KHz}$	75		ms
$t_{set}^*$	Level setting time (see fig. 17)		300		ms
$t_{rec}^*$	Recovery time (see fig. 17)	$\Delta V_i = -40 \text{ dB}$ $f = 1 \text{ KHz}$	150		sec.
$t_{on}^*$	Switch-on-time	$V_o = 1V$	500		ms
$\frac{S+N}{N}$ ***	Signal to noise ratio with ALC	$V_o = 1V$ $R_g = 470\Omega$	64		dB

\* This value depends on external network.

\*\* When the DIN 45511 norm for frequency response is not mandatory the equalization peak at 15 KHz can be avoided - so halving the output noise.

\*\*\* Weighted noise measurement (DIN 45405).



Fig. 11 - Frequency response for the circuit in fig.9 (playback)

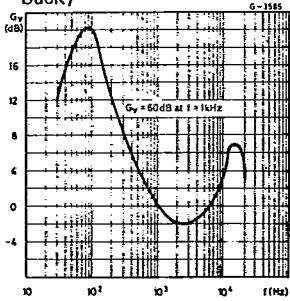


Fig. 12 - Distortion vs. frequency for the circuit in fig. 9 (playback)

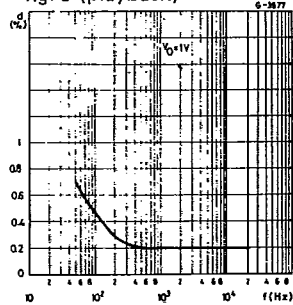


Fig. 13 - Frequency response for the circuit in fig. 9 (recording)

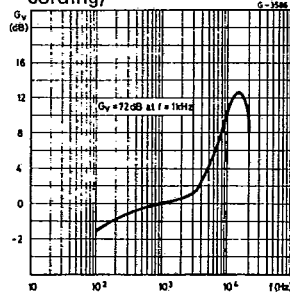


Fig. 14 - Output voltage variation with ALC vs. input voltage for the circuit in fig. 9 (recording)

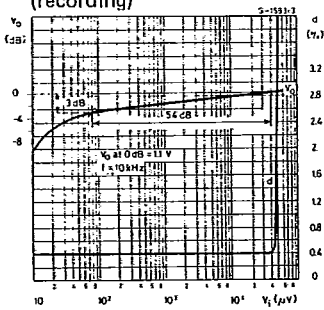


Fig. 15 - Distortion vs. frequency with ALC for the circuit in fig. 9 (recording)

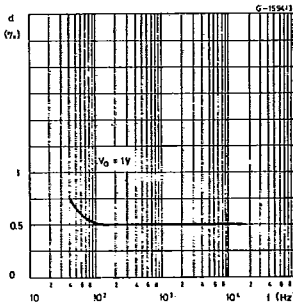


Fig. 16 - Limiting and level setting time vs. input signal variation

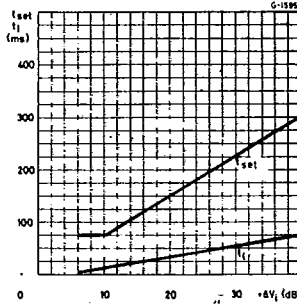


Fig. 17 - Limiting, level setting, recovery time

